

JW5071

28V/2A

Sync. Step-Down Converter

Preliminary Specifications Subject to Change without Notice

DESCRIPTION

The JW[®]5071 is a current mode monolithic buck voltage converter. Operating with an input range of 4.5V-28V, the JW5071 delivers 2A of continuous output current with two integrated N-Channel MOSFETs. At light loads, regulators operate in low frequency to maintain high efficiency.

The JW5071 guarantees robustness with short circuit protection, thermal protection, current run-away protection, and input under voltage lockout.

The JW5071 is available in a 6-pin SOT23-6 package, which provides a compact solution with minimal external components.

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FEATURES

- 4.5V to 28V operating input range 2A output current
- Up to 93% efficiency
- High efficiency at light load
- 600kHz switching frequency
- Input under voltage lockout
- Start-up current run-away protection
- Over current protection and hiccup
- Thermal protection
- Available in SOT23-6 package

APPLICATIONS

- Distributed Power Systems
- Networking Systems
- FPGA, DSP, ASIC Power Supplies
- Green Electronics/ Appliances
- Notebook Computers

TYPICAL APPLICATION



ORDER INFORMATION

	DEVICE ¹⁾	PACKAGE	TOP MARKING ²⁾
		SOT22 6	JMDA
	JW20712010#1KPBF	30123-0	YW□□□
Notes:			
JW # TRPBF Tape and Reel(If" TR" is not shown, it means tube) Package Code Part No.			
JW 000 2) Line1:	Internal control code YW Product code Line2: Joulwatt LOGO	DD Lot number Week code Year code	

PIN CONFIGURATION



ABSOLUTE MAXIMUM RATING¹⁾

VIN, EN Pin	-0.3V to 32V
SW	
BST Pin	SW-0.3V to SW+5.5V
All other Pins	-0.3V to 6V
Junction Temperature ²⁾	150°C
Lead Temperature	
Storage Temperature	65 °C to +150 °C

RECOMMENDED OPERATING CONDITIONS³⁾

Input Voltage V _{IN}	4.5V to 28V
Output Voltage Vout	0.6V to V _{IN} *Dmax
Operating Junction Temperature	40 °C to +125 °C

THERMAL PERFORMANCE⁴⁾

Note:

1) Exceeding these ratings may damage the device. These stress ratings do not imply function operation of the device at any other conditions beyond those indicated under RECOMMENDED OPERATING CONDITIONS.

- 2) The JW5071 includes thermal protection that is intended to protect the device in overload conditions. Continuous operation over the specified absolute maximum operating junction temperature may damage the device.
- 3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.

 θ_{JA}

 θ_{Jc}

ELECTRICAL CHARACTERISTICS

V_{IN} =12V, T_A =25 °C, unless otherwise stated.						
Item	Symbol	Conditions	Min.	Тур.	Max.	Unit
VIN Under Voltage Lock-out Threshold	V _{IN_MIN}	V _{IN} rising	3.95	4.2	4.45	V
VIN Under Voltage Lock-out Hysteresis	VIN_MIN_HYST			400		mV
Shutdown Supply Current	Isd	V _{EN} =0V			3	μA
Supply Current	ΙQ	V _{EN} =5V, V _{FB} =1.2V		110	130	μA
Feedback Voltage	Vfb	-40°C≤Tj≤125°C	588	600	612	mV
Top Switch Resistance	Rds(on)t			90		mΩ
Bottom Switch Resistance	R _{DS(ON)B}			78		mΩ
Top Switch Leakage Current	ILEAK_TOP	V _{IN} =28V, V _{EN} =0V, V _{SW} =0V		0.1	1	μΑ
Bottom Switch Leakage Current	ILEAK_BOT	V _{IN} =28V, V _{EN} =0V, V _{SW} =28V		0.1	1	μΑ
Top Switch Current Limit	ILIM_TOP	Minimum Duty Cycle	3	4	5	A
Switching Frequency	Fsw		450	600	750	kHz
Minimum On Time ⁵⁾	Ton_min			100		ns
Minimum Off Time	T _{OFF_MIN}	V _{FB} =0.4V		120		ns
EN Rising Threshold	Ven_h	V _{EN} rising		1.2	1.35	V
EN Falling Threshold	V _{EN_L}	V _{EN} falling	1.0	1.1		V
EN Hysteresis	Ven_hyst			100		mV
Soft-Start Time ⁵⁾	tss			1		ms
Thermal Shutdown ⁵⁾	T _{TSD}			150		°C
Thermal Shutdown Hysteresis ⁵⁾	T _{TSD_HYST}			20		°C

Note:

5) Guaranteed by design.

PIN DESCRIPTION

SOT23-6 Pin	Name	Description
1	GND	Ground.
2	SW	SW is the switching node that supplies power to the output. Connect the output LC filter from
		SW to the output load.
3	VIN	Input voltage pin. VIN supplies power to the IC. Connect a 4.5V to 28V supply to VIN and
		bypass VIN to GND with a suitably large capacitor to eliminate noise on the input to the IC.
4	FB	Output feedback pin. FB senses the output voltage and is regulated by the control loop to
		0.6V. Connect a resistive divider at FB.
5	EN	Drive EN pin high to turn on the regulator and low to turn off the regulator.
6	BST	Bootstrap pin for top switch. A 0.1uF or larger capacitor and a 10Ω resistor should be
		connected between this pin and the SW pin to supply current to the top switch and top
		switch driver.

BLOCK DIAGRAM



TYPICAL PERFORMANCE CHARACTERISTICS

V_{IN} =12V, V_{OUT} = 3.3V, L = 4.7µH, C_{OUT} = 22µF, T_A = +25°C, unless otherwise noted

Steady State Test

V_{IN}=12V, V_{OUT}=3.3V I_{OUT}=2A



Startup through Enable V_{IN}=12V, V_{OUT}=3.3V I_{OUT}=2A (Resistive load)

Medium Load Operation

0.2A LOAD

Vin

10V/div

Vo(AC)

20mV/div

SW

10V/div

IL

2A/div



Shutdown through Enable

V_{IN}=12V, V_{OUT}=3.3V I_{OUT}=2A (Resistive load)



Heavy Load Operation

2A LOAD



Short Circuit Protection

 V_{IN} =12V, V_{OUT} =3.3V I_{OUT}=2A - Short



Short Circuit Recovery

V_{IN}=12V, V_{OUT}=3.3V I_{OUT}= Short - 2A



2us/div

Light Load Operation 0 A LOAD



Load Transient 1A LOAD \rightarrow 2A LOAD \rightarrow 1A LOAD



TYPICAL PERFORMANCE CHARACTERISTICS





FUNCTIONAL DESCRIPTION

The JW5071 is a synchronous, current-mode, step-down regulator. It regulates input voltages from 4.5V to 28V down to an output voltage as low as 0.6V, and is capable of supplying up to 2A of load current.

Current-Mode Control

The JW5071 utilizes current-mode control to regulate the FB voltage. Voltage at the FB pin is regulated at 0.6V so that by connecting an appropriate resistive divider between V_{OUT} and GND, designed output voltage can be achieved.

PFM Mode

The JW5071 operates in PFM mode at light load. In PFM mode, switch frequency decreases when load current drops to boost power efficiency at light load by reducing switch-loss, while switch frequency increases when load current rises, minimizing output voltage ripples.

Internal Soft-Start

Soft-Start makes output voltage rising smoothly follow an internal SS voltage until SS voltage is higher than the internal reference voltage. It can prevent overshoot of output voltage when startup.

Power Switch

N-Channel MOSFET switches are integrated on the JW5071 to down convert the input voltage to the regulated output voltage. Since the top MOSFET needs a gate voltage greater than the input voltage, a boost capacitor connected between BST and SW pins is required to drive the gate of the top switch. The boost capacitor is charged by the internal 3.7V rail when SW is low.

VIN Under-Voltage Protection

A resistive divider can be connected between V_{IN} and GND, with the central tap connected to EN, so that when V_{IN} drops to the pre-set value, EN drops below 1.1V to trigger input under voltage lockout protection.

Output Current Run-Away Protection

At start-up, due to the high voltage at input and low voltage at output, current inertia of the output inductance can be easily built up, resulting in a large start-up output current. A valley current limit is designed in the JW5071 so that only when output current drops below the valley current limit can the top power switch be turned on. By such control mechanism, the output current at start-up is well controlled.

Over Current Protection and Hiccup

JW5071 has a cycle-by-cycle current limit. When the inductor current triggers current limit, JW5071 enters hiccup mode and periodically restart the chip.

JW5071 will exit hiccup mode while not triggering current limit.

Thermal Protection

When the temperature of the JW5071 rises above 150°C, it is forced into thermal shut-down.

Only when core temperature drops below 130°C can the regulator becomes active again.

APPLICATION INFORMATION

Output Voltage Set

The output voltage is determined by the resistor divider connected at the FB pin, and the voltage ratio is:

$$V_{FB} = V_{OUT} \cdot \frac{R_2}{R_2 + R_3}$$

where V_{FB} is the feedback voltage and V_{OUT} is the output voltage.

If R_2 is determined, and then R_3 can be calculated by:

$$R_3 = R_2 \cdot (\frac{V_{OUT}}{0.6} - 1)$$

For better noise immunity, when input voltage is high (e.g. above 18V), the value of low side divider resistor R_2 is strongly recommended to be no larger than $3k\Omega$.



Input Capacitor

The input capacitor is used to supply the AC input current to the step-down converter and maintaining the DC input voltage. The ripple current through the input capacitor can be calculated by:

$$I_{C1} = I_{LOAD} \cdot \sqrt{\frac{V_{OUT}}{V_{IN}} \cdot \left(1 - \frac{V_{OUT}}{V_{IN}}\right)}$$

where I_{LOAD} is the load current, V_{OUT} is the output voltage, V_{IN} is the input voltage.

Thus the input capacitor can be calculated by the following equation when the input ripple voltage is determined.

$$C_1 = \frac{I_{LOAD}}{f_s \cdot \Delta V_{IN}} \cdot \frac{V_{OUT}}{V_{IN}} \cdot (1 - \frac{V_{OUT}}{V_{IN}})$$

where C_1 is the input capacitance value, fs is the switching frequency, ΔV_{IN} is the input ripple voltage.

The input capacitor can be electrolytic, tantalum or ceramic. To minimizing the potential noise, a small X5R or X7R ceramic capacitor, i.e. 0.1μ F, should be placed as close to the IC as possible when using electrolytic capacitors.

A 22µF/25V ceramic capacitor is recommended in typical application.

Output Capacitor

The output capacitor is required to maintain the DC output voltage, and the capacitance value determines the output ripple voltage. The output voltage ripple can be calculated by:

$$\Delta V_{OUT} = \frac{V_{OUT}}{f_s \cdot L} \cdot (1 - \frac{V_{OUT}}{V_{IN}}) \cdot (R_{ESR} + \frac{1}{8 \cdot f_s \cdot C_{OUT}})$$

where C_{OUT} (C_3 // C_4) is the output capacitance value and R_{ESR} is the equivalent series resistance value of the output capacitor.

The output capacitor can be low ESR electrolytic, tantalum or ceramic, which lower ESR capacitors get lower output ripple voltage.

The output capacitors also affect the system stability and transient response, and a 22μ F~44 μ F ceramic capacitor is recommended in typical application.

Inductor

The inductor is used to supply constant current to the output load, and the value determines the ripple current which affect the efficiency and the

JW5071

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output voltage ripple. The ripple current is typically allowed to be 40% of the maximum switch current limit, thus the inductance value can be calculated by:

$$L = \frac{V_{OUT}}{f_s \cdot \Delta I_L} \cdot (1 - \frac{V_{OUT}}{V_{IN}})$$

where V_{IN} is the input voltage, V_{OUT} is the output voltage, fs is the switching frequency, and ΔI_L is the peak-to-peak inductor ripple current.

External Bootstrap Capacitor

The bootstrap capacitor is required to supply voltage to the top switch driver. A 0.1μ F low ESR ceramic capacitor and a 10Ω resistor are recommended to connected to the BST pin and SW pin.

PCB Layout Note

For minimum noise problem and best operating performance, the PCB is preferred to following the guidelines as reference.

- Place the input decoupling capacitor as close to JW5071 (VIN pin and GND pin) as possible to eliminate noise at the input pin. The loop area formed by input capacitor and GND must be minimized.
- 2. Put the feedback trace as far away from the inductor and noisy power traces as possible.
- 3. The ground plane on the PCB should be as large as possible for better heat dissipation.



Top Layer



Bottom Layer Figure 2. PCB Layout Recommendation

REFERENCE DESIGN

Reference 1:



Figure 3: VIN=5V, VOUT=1.2V/2A

Reference 2:



Figure 4: VIN=12V, VOUT=1.2V/2A

Reference 3:



Figure 5: VIN=12V, VOUT=3.3V/2A

Reference 4:



Figure 6: V_{IN}=24V, V_{OUT}=3.3V/2A

Reference 5:



Figure 7: VIN=12V, VOUT=5V/2A

Reference 6:



Figure 8: VIN=24V, VOUT=5V/2A

TAPE AND REEL INFORMATION





PACKAGE OUTLINE



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